

[54] EXTERNAL CONSTANT SPECIFICATION IN A DIGITAL ELECTRONIC SYSTEM

[75] Inventors: James Russo; Michael R. Levine; Victor Rigotti, all of Ann Arbor; Nicholas Skogler, Ypsilant, all of Mich.

[73] Assignee: Honeywell Incorporated, Minneapolis, Minn.

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[58] Field of Search 364/556, 557; 307/116, 307/117, 246; 374/170, 178, 185; 323/369

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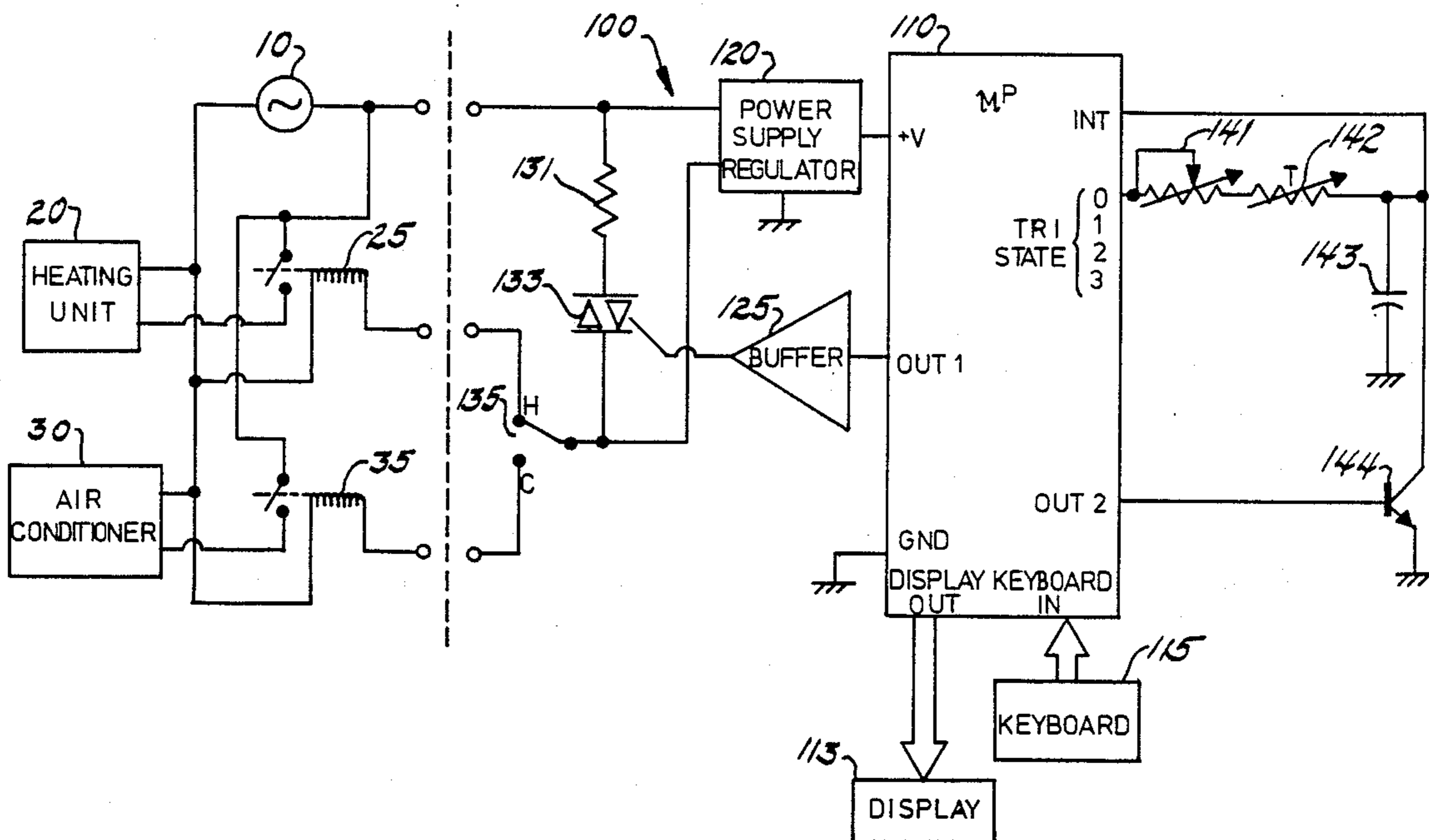
Primary Examiner—Parshotam S. Lall

Assistant Examiner—Ellis B. Ramirez
Attorney, Agent, or Firm—Krass & Young

[57] ABSTRACT

The present invention is a method for specifying a constant in a digital electronic system. This technique employs the ratio of resistances between a reference resistor and a constant specifying resistor. The time constant of the constant specifying resistor and a capacitor is measured by discharging the capacitor, charging the capacitor through the constant specifying resistor and counting the number of predetermined time intervals required for the voltage across the capacitor to reach a predetermined threshold voltage. A similar time constant measurement is made for the reference resistor and the capacitor. The same conditions are employed as far as possible during these two measurements to insure the only differences are in the value of the resistances. The externally specified constant is computed from the digital ratio of the time constant of the constant specifying resistor and the time constant of the reference resistor. This technique is particularly adapted for use in a microprocessor system having several tristate output lines.

4 Claims, 6 Drawing Sheets



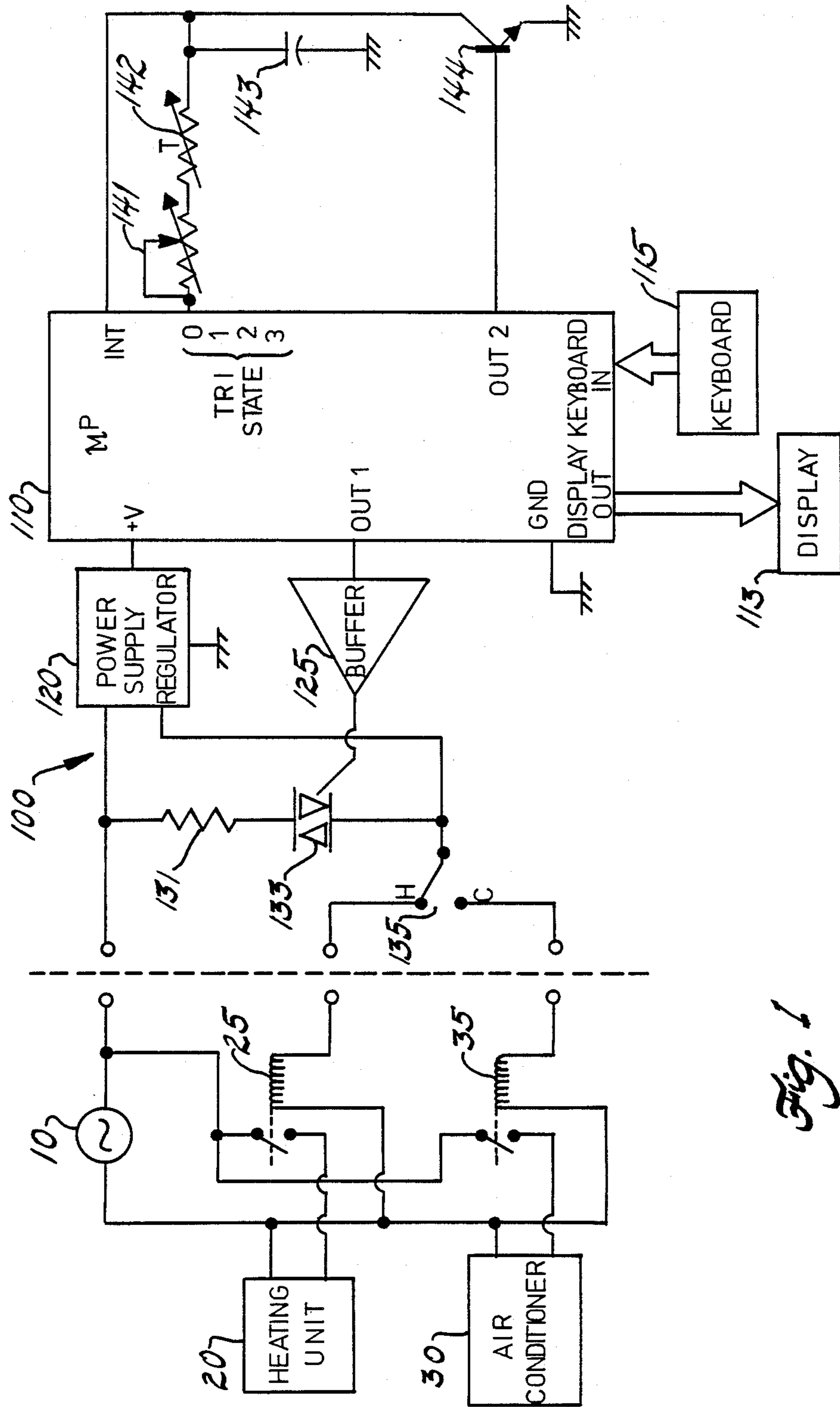
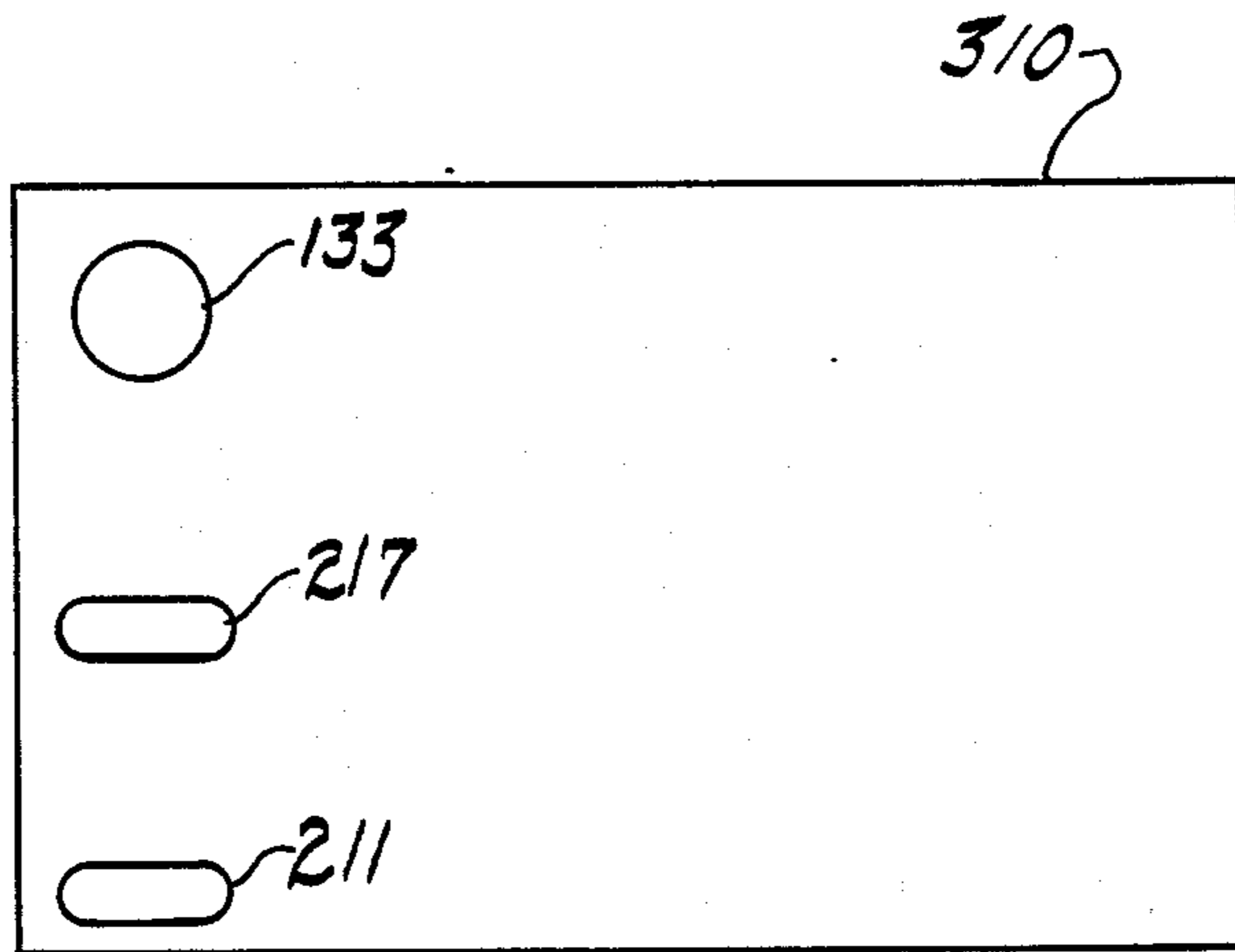
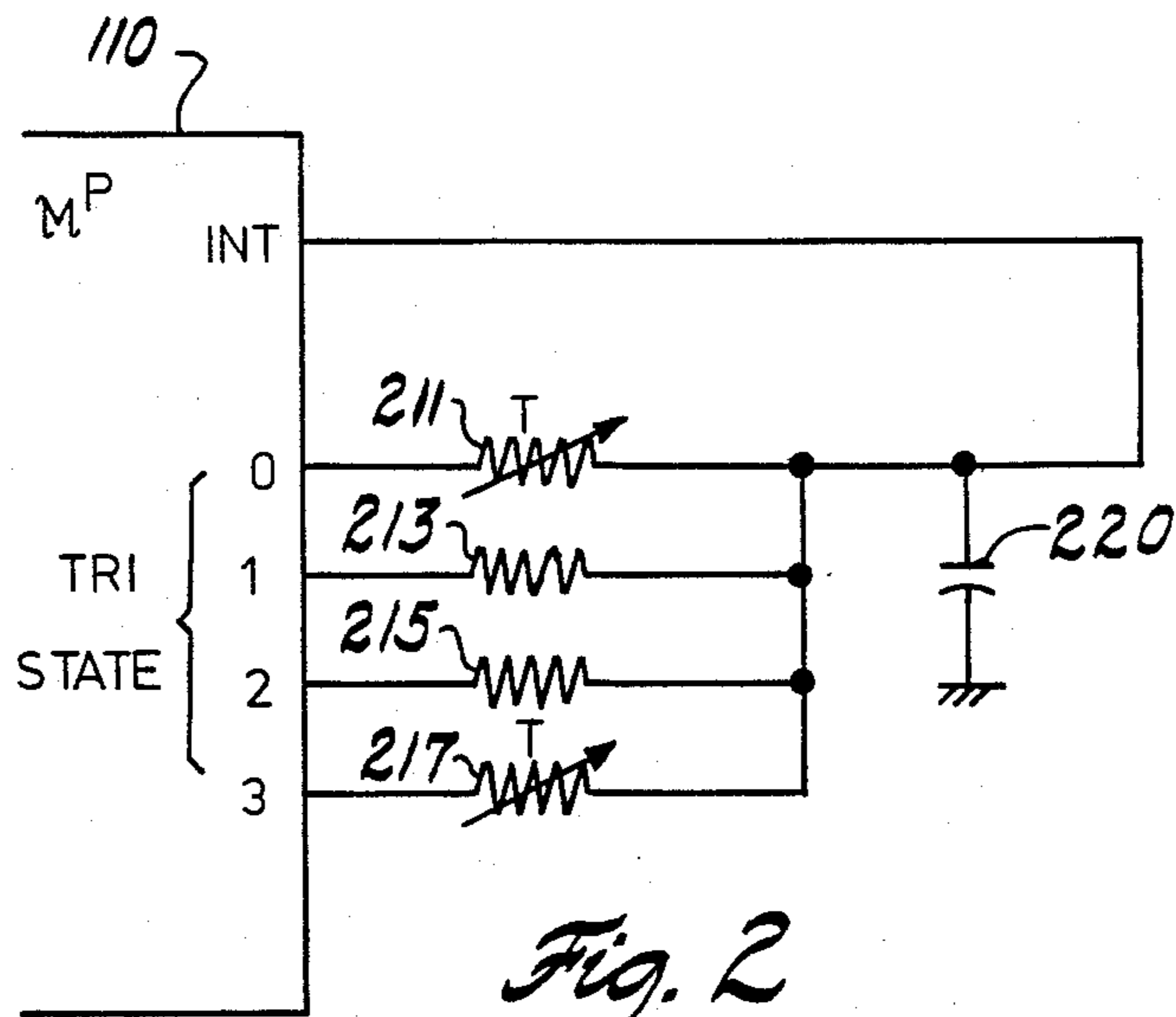
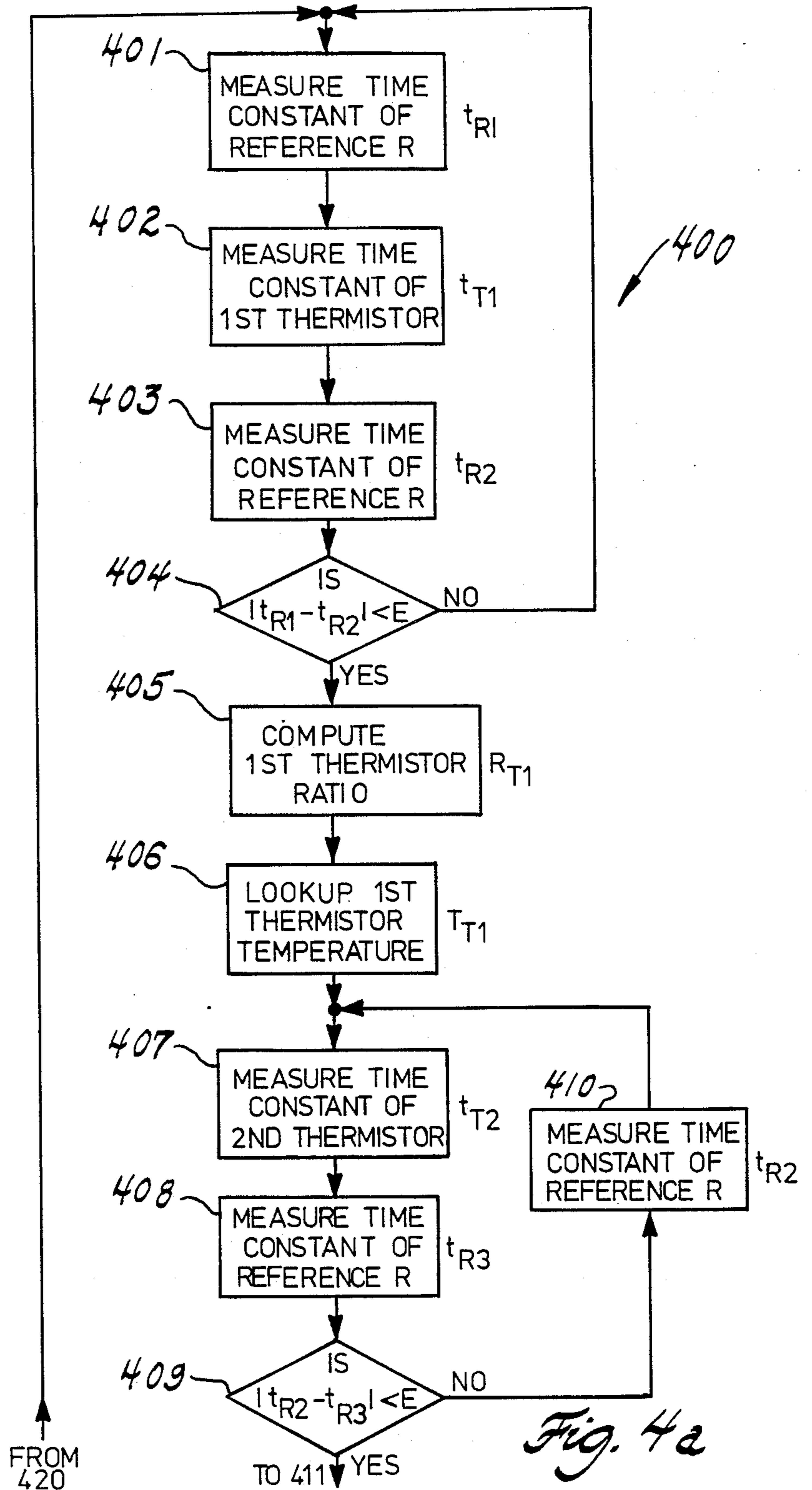


Fig. 1





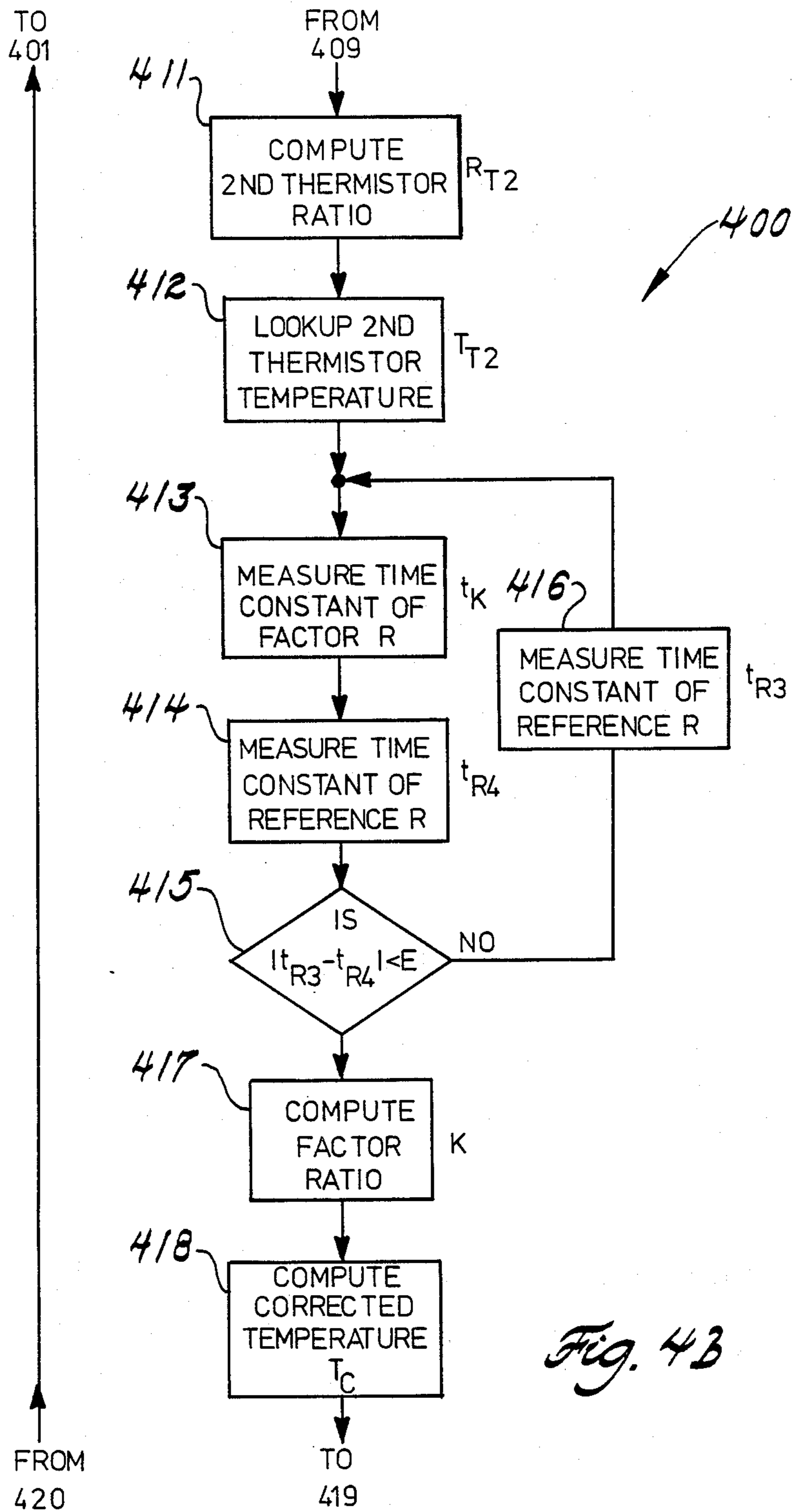


Fig. 4B

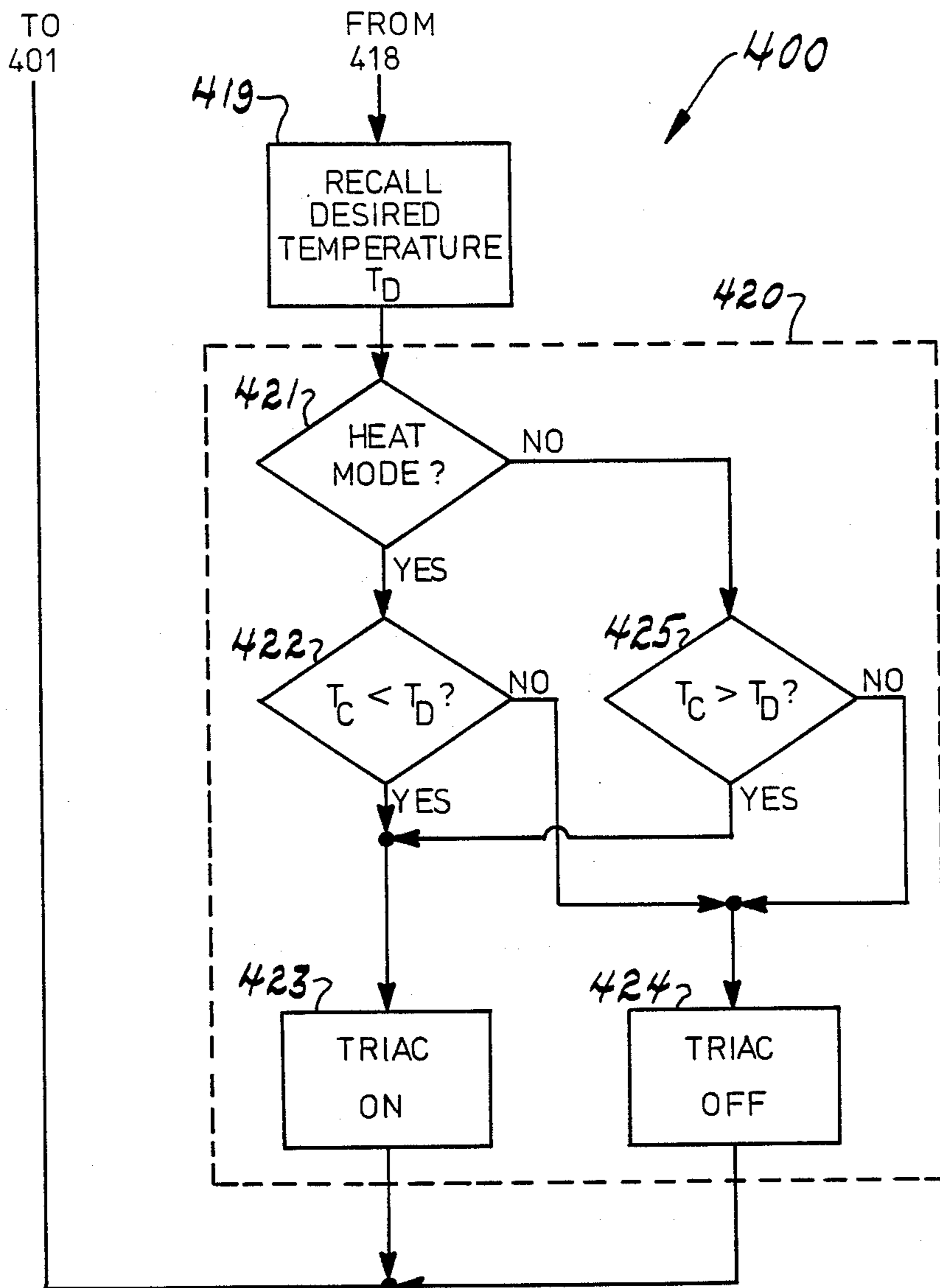


Fig. 40

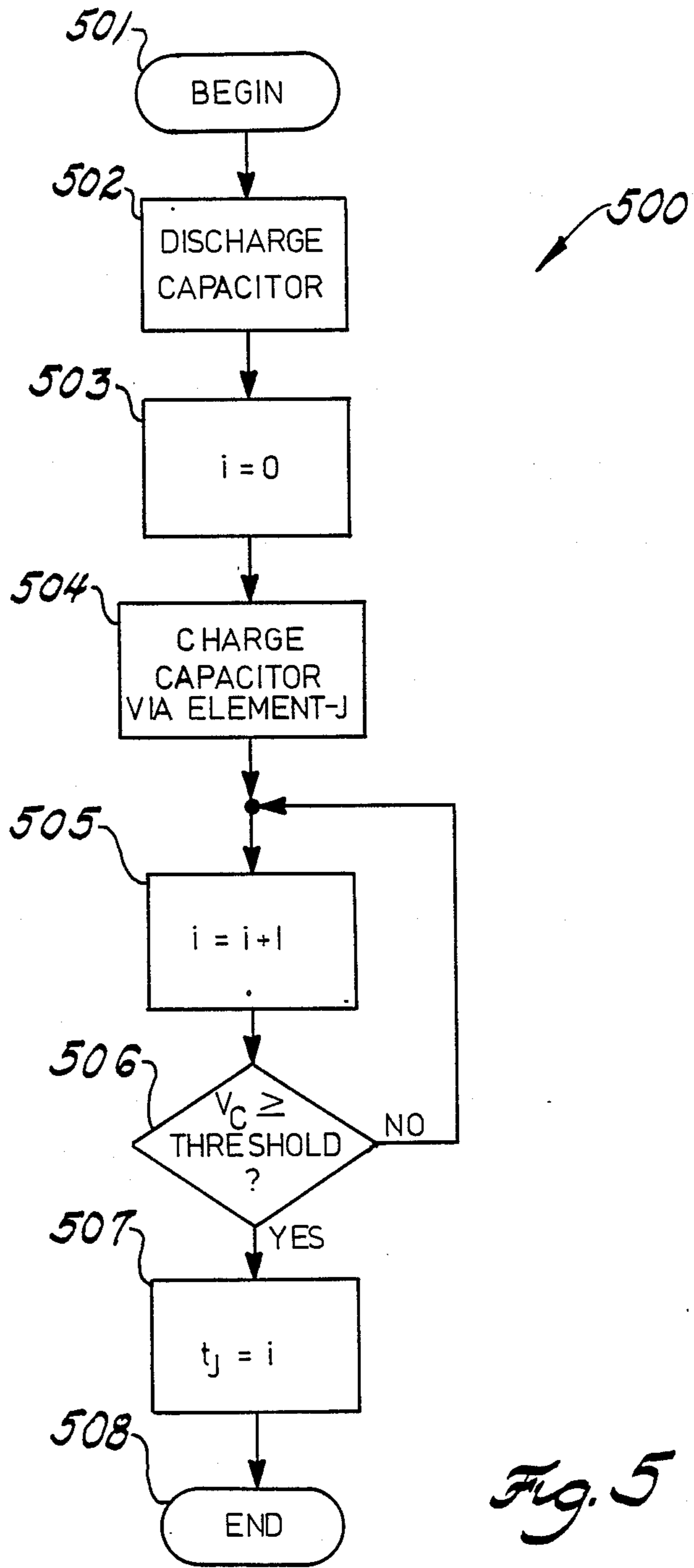


Fig. 5

EXTERNAL CONSTANT SPECIFICATION IN A DIGITAL ELECTRONIC SYSTEM

FIELD OF THE INVENTION

The field of the present invention is the field of providing external specification of constants to a digital electronic system, particularly to a microprocessor system.

BACKGROUND OF THE INVENTION

There is often a need to externally specify a constant to a digital electronic system. In particular, empirically determined parameters or user modifiable features used in specific systems often employ externally specified constants. These constants are employed to modify the operation of the system in particular ways.

There are two main methods used in the prior art to provide an external specification of constants to a digital electronic system. The first of these involves a set of user activated switches. These switches could be used manipulated to indicate a set of one or more binary digits. In accordance with the prior art these switches are often provided in dual in line packages which may be mounted on circuit boards in the same manner as integrated circuits. This packaging technique makes such switches easy to incorporate into a digital electronic system. In addition, these sets of switches are relatively inexpensive. There are disadvantageous features to this technique. Firstly, it is often very difficult for the user to manipulate these switches. Particularly if a number of such switches are mounted in a dual in line package, they are very small. A small instrument, such as a pencil point, is often needed to manipulate these switches. The number of connections required to couple such switches to the digital electronic system is directly proportional to the number of bits required. Many digital electronic systems, particularly low cost microprocessor systems, have only a limited number of connections for all interfacing and thus do not have many connections to devote to such sets of switches.

A second technique employed in the prior art involves the use of a read only memory or a programmable read only memory. These memories can store large amounts of digital data, and thus can be used to provide external specification of a number of constants. These memories differ chiefly in their manner of programming. Read only memories have the data fixed during their manufacture and cannot be changed once constructed. Programmable read only memories are constructed capable of accepting later programming. Some types of programmable read only memories are erasable and capable of being reprogrammed while other types are not. These memories are generally used in the same manner. They are connected to the digital electronic system to an address bus, which enables the digital electronic system to specify which data to be read, and a data bus, which permits the data to be transmitted from the memory to the digital electronic system. This type of interface is more complex than that described above and is more expensive to construct. Because of this and because such memories are capable of specifying a large amount of data, such a system is generally economical only when a small amount of data is to be specified.

It is therefore a need in the industry to provide a low cost method of specifying an external constant of a

limited number of bits which does not require a large number of connections to the digital electronic system.

SUMMARY OF THE INVENTION

The technique of the present invention uses the ratio of two fixed resistors to specify an external constant to a digital electronic system. The relative resistances of these resistors is measured by separately measuring the time constant of an RC circuit including the separate resistors and a common capacitor. This ratio is the externally specified constant.

A measurement sequence is employed to minimize the errors in measurement. Firstly, the time constant of a reference resistor and a capacitor is measured. This is done by discharging the capacitor, charging the capacitor through the reference resistance and counting the number of time intervals of predetermined length required for the voltage across the capacitor to reach a predetermined level. This same process then takes place using a constant specifying resistor. The resistance of this constant specifying resistor is selected to have a ratio relative to the resistance of the reference resistor equal to the constant to be specified. The second measurement takes place under the same conditions by employing the same capacitor, the same voltage supply and the same detector for the predetermined voltage the time constant of the reference resistor and the capacitor is measured a second time. If the two measurements of the time constant of the reference resistor and the capacitor differ by more than a predetermined amount then the measurement sequence is repeated. If the two measurements do not so differ, then the initial count for the time constant of the reference resistor and the capacitor is confirmed. The constant is determined by the digital electronic system by digitally forming the ratio between the two counts. By using the same conditions sources of error are canceled out when the ratio is taken.

This system can be practiced advantageously using a microprocessor having tristate output lines. The reference resistor and the constant specifying resistor are connected at one terminal to separate tristate output lines. Their other terminals are connected to one terminal of the capacitor. The other terminal of the capacitor is connected to a reference voltage. The capacitor may be discharged by generating a "0" output on both tristate output lines. While the time constant of one of the resistances is being measured, the corresponding tristate output line is driven to generate a "1" output while the other tristate output line is driven to the indefinite, high impedance state.

This technique uses a minimum number of connections to the digital electronic system. In particular, several constant specifying resistors can be employed with the same reference resistor. It has been found that with this system accuracies up to twelve bits can be achieved, thus enabling a single constant specifying resistor to specify up to twelve bits.

BRIEF DESCRIPTION OF THE DRAWINGS

These and other aspects of the present invention will become clear from study of the drawings in which:

FIG. 1 illustrates an example of an electronic thermostat employing a microprocessor in accordance with the prior art;

FIG. 2 illustrates connections to the microprocessor for reading and correcting the temperature from a pair of thermistors;

FIG. 3 illustrates the preferred mounting of the triac and the first and second thermistors on a printed circuit board in accordance with the present invention;

FIG. 4 illustrates a flow chart of the self heating correction electronic thermostat employing the circuitry illustrated in FIG. 2; and

FIG. 5 illustrates an example of the subroutine for measuring the time constant of an RC circuit connected in the manner illustrated in FIG. 2.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENT

FIG. 1 illustrates the typical construction of an electronic programmable thermostat in accordance with the prior art. Electronic programmable thermostat 100 includes microprocessor unit 110 which is coupled to display 113 and keyboard 115, power supply regulator 120, a buffer 125 which drives a triac 133, a series resistor 131, a heat/cool mode switch 135 and a temperature sensing circuit including variable resistor 141, thermistor 142, capacitor 143 and transistor 144. Electronic thermostat 100 is connected to a combined heating and air conditioning plant including AC power supply 10, heating unit 20 with an associated relay 25 and air conditioner 30 with associated relay 35.

Electronic thermostat 100 is programmed via keyboard 115 to store a sequence of desired temperatures at desired times. The ambient temperature of the thermostat is measured via thermistor 142. The resistance of thermistor 142 is measured by measuring the time constant of the RC circuit including variable resistor 141, thermistor 142 and capacitor 143. In accordance with the relationship with the measured temperature to the desired temperature, microprocessor unit 110 generates an output at output 1 to buffer 125 which controls whether triac 133 is triggered ON or not. If triac 133 is triggered ON then one of the relays 25 or 35 is supplied with electric power from AC source 10 depending upon the state of heat/cool mode switch 135. The actuated relay in turn operates the corresponding heating unit 20 or air conditioner 30.

Power supply regulator 120 is connected to receive power from the series connection of resistor 131 and triac 133. This power comes from AC power source 10. In the event that triac 133 is not triggered ON, then the full voltage of the AC power source 10 appears across the input to power supply regulator 120. If, on the other hand, triac 133 is triggered ON, then the voltage input to power supply regulator 120 is the voltage across triac 133 together with the IR voltage appearing across resistor 131. As a result, the input voltage applied to power supply regulator 120 varies widely depending on whether or not triac 133 is triggered ON. Therefore, power supply regulator 120 is employed to smooth these wide variations in supply voltage to derive the appropriate voltage for driving the microprocessor unit 110.

The ambient temperature is measured by the microprocessor unit 110 from the variable resistor 141, thermistor 142, capacitor 143 and transistor 144. Variable resistor 141 and thermistor 142 together with capacitor 143 form a RC time constant circuit. The resistance of thermistor 142 varies as a function of ambient temperature. This resistance is measured by measuring the time constant of this RC circuit. Microprocessor unit 110 includes a plurality of tristate output lines 0-3. One of these tristate output lines is applied to one end of the RC

time constant circuit, while the other end of this RC time constant circuit is connected to ground.

These tristate output lines of microprocessor unit 110 have unique characteristics. The output of each of these tristate output lines can be driven to the power supply voltage in response to a "1" output or to the ground voltage in response to a "0" output. In this state the output line is connected through a relatively small resistance of 50 to 100 ohms to the respective voltage source. It is also possible to drive any of these tristate output lines to an indeterminate state in which the output line is connected to neither the positive supply voltage or ground. Typically, the isolation between the output and either the power supply voltage or ground is greater than 5 megohms when in the indeterminate state.

Transistor 144 is connected across capacitor 143. Transistor 144 has its base connected to output 2 of the microprocessor unit 110. Transistor 144 is employed to discharge any charge which is stored on capacitor 143. In response to a "1" output on output 2 from microprocessor unit 110, transistor 144 is biased ON and discharges any charge stored in capacitor 143. In response to a "0" output at output 2 of microprocessor unit 110, transistor 144 is biased OFF and does not effect the charge on capacitor 143.

Lastly, the output voltage across capacitor 143 is connected to an interrupt input of microprocessor unit 110. This interrupt input is employed to signal microprocessor unit 110 that the voltage across capacitor 143 has exceeded this predetermined value.

During operation of electronic thermostat 100, the temperature indicated by thermistor 142 is measured employing a program stored within microprocessor unit 110 to control the output supplied to this measurement circuit. Firstly, transistor 144 is biased ON in order to discharge any charge stored in capacitor 143. During this operation the tristate output 0 could be held at either the indeterminate state or at ground through the output of a digital "0". After transistor 144 has been biased ON for a sufficient period to substantially discharge capacitor 143, output 2 supplied to the base of transistor 144 is changed to bias this transistor to OFF.

The time constant of the RC circuit is then measured. The tristate output line 0 is driven to the power supply voltage via the output of a digital "1". At the same time a timer circuit is initialized and begins to accumulate the elapsed time. This output from the tristate output line 0 causes the power supply voltage to be applied across variable resistor 141, thermistor 142 and capacitor 143. As a consequence, a current flows through this RC circuit and begins accumulate charge within capacitor 143. This accumulating charge within capacitor 143 causes the voltage there across to increase. This voltage is sensed by the interrupt input INT of microprocessor unit 110. When the voltage across capacitor 143 exceeds the predetermined threshold of interrupt input INT, this input is triggered. Microprocessor unit 110 is programmed to stop the elapsed time clock at this time. As a consequence, the time then indicated by the elapsed time clock is a measure of the time constant of the RC circuit. Ideally, the only variable in this time constant measurement is the resistance of thermistor 142. Therefore, this measured elapsed time is a measure of the temperature sensed by thermistor 142. Typically, microprocessor unit 110 includes a look-up table to convert this time to the corresponding temperature.

There are several disadvantages with the foregoing manner for sensing the temperature. Typically, the resistance of the thermistor is not the only variable in the system. In particular, the capacitance of capacitor 143 may change slowly with time. For this reason, variable resistor 141 has been included in the RC time constant circuit in order to adjust for these factors. Variable resistor 141 may be eliminated if a high precision thermistor 142 and a high precision capacitor 143 are employed. However, the magnitude of the supply voltage is an additional variable. As explained above, the supply voltage applied to microprocessor unit 110 is typically not very well regulated. Therefore, the voltage used to charge capacitor 143 through variable resistor 141 and thermistor 142 can vary depending upon the state of triac 133. This variation in supply voltage causes a variation in the time required to charge capacitor 143 to the predetermined voltage. These variations in the supply voltage can cause inaccuracies in the measured temperature.

As a result of these factors, the designer in accordance with the prior art could make one of two relatively disadvantageous choices. Firstly, a highly complex power supply regulator 120 which provides an excellent regulation of the power supply to microprocessor unit 110 could be employed together with a high precision thermistor 142 and a high precision capacitor 143. A design of this type would serve to eliminate many of the sources of inaccuracy in the measurement of a temperature in accordance with the above described circuit. On the other hand, this provision for a highly accurate power supply regulator 120 and precision components would require additional expense in the construction of the electronic thermostat. On the other hand, a relatively low accuracy power supply regulator 120 could be employed with relatively inexpensive but less precise components. This design technique reduces the cost of electronic thermostat 110 with a consequent decrease in the accuracy of the temperature measurement and thus the accuracy of the control of temperature.

The prior art thermostat illustrated in FIG. 1 has a problem regarding the self-heating of the temperature sensing thermistor by the triac employed for switching the electric power. As triac 133 carries the current for actuation of either relay 25 or relay 35, the current through this element heats it up. In many cases electronic thermostat 100 is unstable because the heating of the triac 133 causes thermistor 142 to be warmed. Even if instability is avoided, this self-heating causes inaccurate measurement of the temperature, and therefore inaccurate thermostatic control.

Control of the self-heating from the triac includes several problems. Firstly, the triac heats only when it is on. Therefore, greater amounts of heat is generated when operating at a higher duty cycle. Such a higher duty cycle occurs when there is a greater heating load or a greater cooling load therefore requiring the selected heating unit 20 or air conditioner 30 to be energized a greater percentage of the time. This type of self-heating can cause wide variations in temperature during the operation of air conditioning. In particular, if the air conditioner is actuated and the triac causes self-heating of the thermistor, the air conditioner may lock on or be on for extended periods of time trying to overcome self-heating before turning off and then be off for an extended period of time for the temperature to return to the set point.

Secondly, although the voltage across the triac is fixed by the physics of this semiconductor device, the current through this device is not known. The amount of current drawn by the relay circuit controls the amount of current taken by the triac. In typical installations this current can vary from between 100 milliamperes and 1 ampere, a ratio of a factor of 10. In particular if an electronic thermostat is intended for retrofit use with an existing heating and air conditioning plant, it is impossible to estimate the amount of self-heating by the triac due to these variations in the current drawn by the relay.

Lastly, the air flow along the wall at which the electronic thermostat is mounted can cause additional problems. In order to minimize self-heating by the triac, the triac is typically mounted at the upper portion of a vertically oriented circuit board with the thermistor disposed on a lower portion of the circuit board. This is an attempt to prevent air heated by the triac from warming and self-heating the thermistor. However, it is generally impossible to determine the direction of air flow along the wall to which the electronic thermostat is mounted. In many installations, the air flow is either upward or transverse and therefore air heated by the triac does not severely heat the thermistor. In other installations the air flow may be downward causing the hot air from the triac to warm the thermistor. Typical instructions for retrofit electronic thermostats suggest moving the location of the thermostat if such problems are encountered. However, in retrofit installations this type of solution is impractical.

FIG. 2 illustrates a circuit for use with a microprocessor based electronic thermostat such as illustrated in FIG. 1 which can correct for the self heating caused by the triac switching element. FIG. 2 illustrates a RC time circuit including a plurality of resistances and a single capacitance, capacitor 220. Thermistor 211 is connected between the tristate output line 0 of microprocessor device 110 and capacitor 220. Similarly, a high precision fixed resistor 213 is connected between tristate output line 1 and the capacitor. Further, a second high precision fixed resistor 215 is connected between tristate output line 2 and capacitor 220. Lastly, a second thermistor 217 is connected between tristate output line 3 and capacitor 220. Each of these resistive elements forms a RC time constant circuit with the capacitor 220 when actuated by the respective tristate output lines 0-3. Although a discharge circuit such as transistor 144 is not illustrated in FIG. 2, it would be understood by those skilled in the art that such a discharge circuit could be employed.

Measurement of the resistance of thermistor 211 occurs in a manner similar to that described above in conjunction with the description of FIG. 1. Capacitor 220 is discharged. This could occur through a discharge device such as transistor 144 illustrated in FIG. 1, or it could occur by tying all of the tristate output lines 0-3 to the ground potential by applying a "0" output to these lines. This latter technique is believed advantageous because it reduces the need for components external to the microprocessor unit 110. Once the capacitor 220 has been sufficiently discharged, an indeterminate high impedance output is applied to tristate output lines 1-3. At the same time, a "1" output is applied to tristate output line 0. An elapsed timer is started in a manner similar to that described above. When the voltage across capacitor 220 reaches the predetermined threshold voltage of the interrupt input INT of microproces-

sor unit 110, then the elapsed time on the elapsed time counter corresponds to the resistance of thermistor 211. This in turn corresponds to the temperature at thermistor 211.

After this measurement has been made then a second measurement is made. Firstly, capacitor 220 is discharged in a manner similar to that described above. Then capacitor 220 is charged via precision fixed resistor 213. This is achieved by applying an indeterminate high impedance output to tristate output lines 0, 2 and 3 and by applying a digital "1" output to tristate line 1. This applies the power supply voltage across the RC circuit including precision fixed resistor 213 and capacitor 220. In the same manner as previously described, the elapsed time is measured from the first application of voltage across capacitor 220 until the voltage across capacitor 220 exceeds the predetermined threshold of the interrupt input INT. This elapsed time is a measure of the resistance of precision fixed resistor 213.

In accordance with the preferred embodiment, the resistance selected for the precision fixed resistor 213 is approximately equal to the nominal value of the resistance of thermistor 211 at a selected reference temperature, such as 70° F. The arithmetic computational capability of the microprocessor 110 is then used to compute the ratio of the time constant measured through thermistor 211 and the time constant measured through precision resistor 213. This ratio of times gives an indication of the ratio of the resistance of thermistor 211 to the resistance of precision fixed resistor 213. The computation of this ratio has the advantageous effect of eliminating sources of variability. The value of capacitance 220 is the same for each measurement. In addition, if these time constant measurements take place sequentially, the power supply voltage of microprocessor 110 cannot change greatly during this interval. Thus the formation of this ratio serves to eliminate many of the sources of inaccuracy of the measurement of the resistance of thermistor 211. This serves to permit a more precise measure of the temperature.

FIG. 3 illustrates the manner of mounting triac 133, first thermistor 211 and a second thermistor 217. Triac 133 is mounted on printed circuit board 310 at an upper portion thereof. In accordance with the typical installation, printed circuit board 310 is vertically mounted parallel to the wall upon which the electronic thermostat is mounted. Triac 133 is preferably mounted at an upper portion thereof in order to reduce the heating of first thermistor 211 via hot air from triac 133. Thermistor 211 is preferably mounted at a lower portion of the printed circuit board 310. The second thermistor 217 is mounted on printed circuit board 310 at a position intermediate between triac 133 and first thermistor 211.

FIG. 2 also illustrates fixed resistor 215 and second thermistor 217. In a manner that will be further described below, the temperature measured by second thermistor 217 is employed to correct for the effect of the heating of triac 133 upon the first thermistor 211. In general the greater the difference between the temperature measured by these two thermistors the greater the correction. The temperature measured by second thermistor is determined in a manner similar to the method used with regard to first thermistor 211. In addition the resistance of the fixed resistor 215 is measured by measuring the time constant of this resistor and the capacitor 220. This takes place in a manner similar to the measurement of the resistance of precision fixed resistor 213. The ratio of the resistances of fixed resistor

215 and precision fixed resistor 213 is employed in the measured temperature correction computation. This will be more fully explained below.

FIG. 4 illustrates program 400 for providing the self-heating correction in accordance with the circuits illustrated in FIG. 3. Program 400 is a continuous loop which repeatedly measures the temperature, corrects the measured temperature and provides the necessary control based upon the relationship between the desired temperature and the measured temperature.

The description of program 400 will begin with processing block 401 which measures the time constant through precision fixed resistor 213. This measure of the time constant takes place in accordance with subroutine 500 illustrated in FIG. 5. Program 400 next measures the time constant through the first thermistor 211 (processing block 402). Then program 400 again measures the time constant through precision fixed resistor 213 (processing block 403).

Program 400 then tests to determine if the two measurements of the time constant of the precision fixed resistor 213 are within a predetermined error limit E (decision block 404). This is accomplished by taking the absolute value of the difference between the first measurement of the time constant t_{R1} and the second measurement of the time constant t_{R2} . If this absolute value is less than the error limit E, then the two measurements are within the predetermined error limit. This test insures that the measurement conditions have not changed significantly during the measurement process. If the two measurements are not within the predetermined error limit, then control of Program 400 passes to processing block 401 to repeat the measurement sequence. Program 400 remains in this loop until the measurements are within the predetermined error limit E.

Program 400 measures the ratio for the first thermistor in accordance with the following formula:

$$R_{T1} = t_{T1} / t_{R1}$$

where R_{T1} is the ratio for the first thermistor, t_{T1} is the time constant for the first thermistor, and t_R is the first measured time constant of the precision fixed resistor 213 (processing block 405). Program 400 then performs a table look-up to determine the temperature T_{T1} measured by the first thermistor (processing block 406).

Program 400 then measures the time constant t_{T2} of the second thermistor 217 (processing block 407). Program 400 measures the time constant through precision fixed resistor 213 for the third time (processing block 408). Program 400 then makes a test similar to that indicated above in relation to the measurement of the time constant of the first thermistor 211. Program 400 tests to determine if the absolute value of the difference between the second measured time constant of the precision fixed resistor t_{R2} and the third measured time constant of the precision fixed resistor t_{R3} is less than a predetermined error limit E (decision block 409). If it is not, then the second measurement of the time constant of precision fixed resistor 213 is repeated (processing block 410) and control passes to processing block 407 to repeat the measurement process. As previously described, Program 400 remains in this loop until the measurement is complete.

The ratio for the second thermistor is then computed (processing block 411). This is performed employing the formula:

$$R_{T2} = t_{T2} / t_{R2}$$

where R_{T2} is the ratio for the second thermistor, t_{T2} is the time constant for the second thermistor, and t_{R2} is the second measured time constant of the precision fixed resistor 213. This ratio value is employed with a look up table in order to determine the temperature T_{T2} measured by the second thermistor (processing block 412).

Program 400 then measures the time constant t_K through constant specifying resistor 215 (processing block 413). As will be explained in further detail below, this resistor is employed to input a constant into microprocessor unit 110. The time constant through precision fixed resistor 213 is measured for the fourth time (processing block 414). Program 400 then makes a test similar to that indicated above in relation to the measurement of the time constant of the thermistor. Program 400 tests to determine if the absolute value of the difference between the third measured time constant t_{R3} of the precision fixed resistor 213 and the fourth measured time constant t_{R4} of the precision fixed resistor 213 is less than a predetermined error limit E (decision block 415). If it is not, then the third time constant t_{R3} of precision fixed resistor 213 is again measured (processing block 416) and control passes to processing block 413 to repeat the measurement process. As previously described, Program 400 remains in this loop until the measurement is complete.

Program 400 then computes a ratio for the fixed resistor 210. This is computed using the following formula:

$$K = t_K / t_{R3}$$

where K is the factor to be computed, t_K is the original measured time constant for fixed resistor 215, and t_{R3} is the third measured time constant of the precision fixed resistor 213 (processing block 417). This factor K is employed in the formula for computation of the corrected temperature. Thus, the value of precision resistor 215 is employed to specify the constant K .

Program 400 then computes the corrected measured temperature T_C (processing block 418). This is computed in accordance with the following formula:

$$T_C = T_{T1} - K * (T_{T2} - T_{T1})$$

The constant K determines the percentage of the difference between the temperature measured by the second thermistor and the temperature measured by the first thermistor that is subtracted from the temperature measured by the first thermistor. In the construction illustrated in FIG. 3, it is expected that generally the temperature measured by the second thermistor 217 will be greater than that measured by the first thermistor 211 because the second thermistor 217 is closer to triac 133 which is a source of heat. The greater the difference in temperature in that measured by the two thermistors, the greater the temperature gradient near the thermistor and the greater the correction factor necessary to correct for the true ambient temperature in the absence of self heating. In the event that the difference in temperature between the temperatures measured by the two thermistors is zero or very small, then the temperature measured by first thermistor 211 is an accurate reflection of the ambient temperature and a small or zero correction factor is indicated.

In general, a less precise second thermistor 217 may be employed in this application. That is, it is not neces-

sary for the temperature versus resistance curve of the second thermistor to be as precisely known prior to installation. The offset between the first and second thermistors can be measured when the triac 133 has been off for a long period of time, insuring that there is no self-heating. This correction factor can be added to the temperature determined by the look-up table (processing block 412) in order to arrive at the temperature at the second thermistor.

The reason why the constant K is specified by resistor 215 rather than being specified by the program of microprocessor unit 110 is as follows. It is believed that the optimum value of K employed in the temperature correction formula above is critically dependent upon the mechanical design of the thermostat. It is proposed that the completed mechanical design of a particular thermostat be tested in an environmental chamber to empirically determine the optimum value of the constant K . If the constant K were stored in the program of microprocessor device 110, then the microprocessor design cannot be complete until the mechanical design is completed and this empirical determination of the constant K is made. By using resistor 215 to specify the constant K , the design of microprocessor unit 110 is not dependent upon the completion of the mechanical design and empirical testing. The specification of constant K takes place by selection of the resistance of resistor 215 independent of the microprocessor program. In addition, the same type of microprocessor unit 110 can be employed in differing mechanical designs with resistor 215 selected according to the optimum constant K for that particular design. Thus, only one type of microprocessor need be manufactured and stocked, decreasing the manufacturing cost of the thermostat.

Program 400 then recalls the desired temperature T_D (processing block 419). In accordance with the preferred embodiment, this desired temperature is recalled from a sequence of desired temperatures for particular times stored within microprocessor unit 110. The manner of storage of these desired temperatures is in accordance with the prior art and will not be further discussed. This recalled desired temperature T_D is employed in the thermostatic control function described below.

Program 400 then performs the thermostatic control function (subroutine 420). As shown in subroutine 420 the program performs a simple on/off function based upon the relationship of the measured ambient temperature T_C and the desired temperature T_D . Those skilled in the art would realize that other thermostatic control functions are feasible.

Subroutine 420 begins by testing whether the thermostat is in heating mode (decision block 421). If this is the case then subroutine 420 tests to determine if the corrected measured temperature T_C is less than the desired temperature T_D . If the corrected measured temperature is less than the desired temperature T_D then triac 133 is triggered ON (processing block 423), otherwise triac 133 is OFF (processing block 424). In the event that the thermostat is not in heating mode, it is in cooling mode. Subroutine 420 tests to determine if the corrected measured temperature T_C is greater than the desired temperature T_D (decision block 425). If this is the case, then triac 133 is triggered ON (processing block 423). Otherwise triac 133 is OFF (decision block 424). In any event control of program 400 returns to processing block 401 to repeat the measurement and control process.

Program 500 illustrated in FIG. 5 is a flow chart of the manner in which the time constants are measured in program 300. Subroutine 500 is employed to embody processing blocks 401, 402, 403, 407, 408, 410, 413, 414 and 416 illustrated in FIG. 4. Subroutine 500 is begun at start block 501. Subroutine 500 first discharges capacitor 220 (processing block 502). This could take place with the use of an external device such as transistor 144 illustrated in FIG. 1 or via the tristate output lines 0-3. Next, an index variable i is initialized (processing block 503). The capacitor 200 is then charged via the particular element J (processing block 504) by actuation of the proper tristate output line 0-3. Subroutine 500 then increments the index variable i (processing block 505). Next, subroutine 500 tests to determine whether or not the voltage across capacitor 220 V_C is greater than or equal to the threshold voltage (decision block 506). If this is not the case then the measurement is not complete and subroutine 500 returns to processing block 505. If this is the case then the measurement is complete. The measured time for the element J t_{rj} is set equal to the index variable i (processing block 507). Subroutine 500 is then complete and is exited via end block 508.

In a practical embodiment of subroutine 500, the incrementing step 505 would be performed by adding 1 to a memory register within microprocessor unit 110. The processing of determining whether or not the voltage across the capacitor V_C exceeds the predetermined threshold of decision block 506 can be employed via the interrupt input of INT microprocessor device 110. Thus for example the incrementing of the index variable i can occur within a closed loop and this loop be broken only at the receipt of the interrupt. The interrupt is detected when the voltage across the capacitor exceeds the predetermined threshold of the interrupt input INT. This interrupt signal stops the incrementing of the index variable i and causes this value to be stored as the measured time. Thus the clock which controls the rate of operation of the microprocessor device 110 serves as a timer to time the number of increments of the index variable i .

We claim:

1. A digital constant specification system in a digital electronic system, said digital constant system comprising:

- a precision fixed resistor having a predetermined reference resistance between first and second terminals;
- a constant specifying fixed resistor having a constant specifying resistance between first and second terminals;
- a capacitor having a first terminal connected to said second terminal of said precision fixed resistor and said second terminal of said constant specifying resistor and a second terminal connected to a predetermined reference voltage;
- a discharge means connected to said capacitor for discharging said capacitor;
- a first time constant measuring means for measuring the reference time constant of said precision fixed resistor by discharging said capacitor via said discharge means, charging said capacitor through said first terminal of said precision fixed resistor and counting the number of predetermined time intervals required for the voltage across said capacitor to reach a predetermined threshold voltage;
- a second time constant measuring means for measuring the constant specifying time constant of said

constant specifying resistor by discharging said capacitor via said discharge means, charging said capacitor through said first terminal of said constant specifying resistor and counting the number of predetermined time intervals required for the voltage across said capacitor to reach a predetermined threshold voltage;

- a measurement sequence means connected to said first time constant measurement means and said second time constant measurement means for producing a sequence of measurements by causing said first time constant measurement means to measure said reference time constant a first time thereby producing a first count, then causing said second time constant measurement means to measure said constant specifying time constant, then causing said first time constant measurement means to measure said reference time constant a second time thereby producing a second count, and for comparing said first count and said second count and repeating said sequence of measurements if said first count and said second count differ by more than a predetermined amount; and
 - a ratio means for computing the digital ratio of the count of predetermined time intervals of said reference time constant and the count of predetermined time intervals of said constant specifying time constant, said ratio being said specified digital constant.
2. A digital constant specification system for use with a microprocessor device having at least two tristate output lines and an interrupt input, said digital constant specification system comprising:
- a precision fixed resistor having a predetermined reference resistance between first and second terminals, said first terminal being connected to a first tristate output line;
 - a constant specifying fixed resistor having a constant specifying resistance between first and second terminals, said first terminal connected to a second tristate output;
 - a capacitor having a first terminal connected to said interrupt input, said second terminal of said precision fixed resistor and said second terminal of said constant specifying resistor, and a second terminal connected to a reference voltage;
 - a discharge means connected to said capacitor and controlled by said microprocessor device for discharging said capacitor; and
 - the microprocessor device being programmed to measure the reference time constant of the reference resistance of said precision fixed resistor and said capacitor by discharging said capacitor via said discharge means, charging said capacitor through said first terminal of said precision fixed resistor by applying a digital "1" output to said first tristate output line and applying a high impedance output to the other of said at least two tristate output lines and counting the number of predetermined time intervals required for the voltage across said capacitor to reach a predetermined threshold voltage at said interrupt input, thereby producing a first count,
 - measure the constant specifying time constant of the constant specifying resistance of said constant specifying fixed resistor and said capacitor by discharging said capacitor via said discharge means, charging said capacitor through said first terminal of said constant specifying resistor by

applying a digital "1" output to said second tri-state output line and applying a high impedance output to the other of said at least two tristate output lines and counting the number of predetermined time intervals required for the voltage across said capacitor to reach a predetermined threshold voltage at said interrupt input, 5

measure the reference time constant of the reference resistance of said precision fixed resistor and said capacitor again by discharging said capacitor via said discharge means, charging said capacitor through said first terminal of said precision fixed resistor by applying a digital "1" output to said first tristate output line and applying a high impedance output to the other of said at least two tristate output lines and counting the number of predetermined time intervals required for the voltage across said capacitor to reach a predetermined threshold voltage of said interrupt input, thereby producing a second count, 10

compare said first count and said second count, if said first count and said second count differs by more than a predetermined amount, set said first count equal to said second count, again measure the constant specifying time constant of the constant specifying resistance of said constant specifying fixed resistor and said capacitor, again measure the reference time constant of the reference resistance of said precision fixed resistor and said capacitor thereby producing a second count, and again compare said first count and said second count, and 15

if said first count and said second count do not differ by more than said predetermined amount compute the digital ratio of said reference time constant and said constant specifying time constant, said digital ratio being said digital constant. 20

3. The digital constant specification system claimed in claim 2, wherein: 25

said microprocessor device is further programmed to count the number of predetermined time intervals required for the voltage across said capacitor to reach a predetermined threshold voltage at said interrupt input by incrementing a memory register at the rate of operation of said microprocessor under closed loop program operation until an interrupt input is received. 30

4. A method of specifying a digital constant comprising sequentially: 35

measuring the reference time constant of a precision fixed resistor and a capacitor by discharging said capacitor, charging said capacitor through said precision fixed resistor and measuring the time for the voltage across said capacitor to reach a predetermined threshold voltage, thereby producing a prior reference time constant measure;

then measuring the constant specifying time constant of a constant specifying fixed resistor and said capacitor by discharging said capacitor, charging said capacitor through said constant specifying fixed resistor and measuring the time for the voltage across said capacitor to reach a predetermined threshold voltage; 40

then again measuring the reference time constant of a precision fixed resistor and said capacitor by discharging said capacitor, charging said capacitor through said precision fixed resistor and measuring the time for the voltage across said capacitor to reach a predetermined threshold voltage, thereby producing a present reference time constant measure;

then comparing said prior reference time constant measure and said present time constant measure; and 45

then, if said prior reference time constant measure differs from said present time constant measure by more than a predetermined amount, setting said prior reference time constant measure equal to said present reference time constant measure, then again measuring the constant specifying time constant of said constant specifying fixed resistor and said capacitor, then again measuring the reference time constant of said precision fixed resistor and said capacitor thereby producing a new present reference time constant measure, and then again comparing said prior reference time constant measure and said present time constant measure; and 50

then, if said prior reference time constant measure does not differ from said present time constant measure by more than said predetermined amount, computing the digital ratio of said constant specifying time constant and said prior reference time constant, said computed digital ratio being the specified digital constant. 55

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UNITED STATES PATENT AND TRADEMARK OFFICE
CERTIFICATE OF CORRECTION

PATENT NO. : 4,829,458
DATED : May 9, 1989
INVENTOR(S) : Russo et al

It is certified that error appears in the above-identified patent and that said Letters Patent is hereby corrected as shown below:

Column 8, line 42, "t_R" should be --t_{R1}--.

**Signed and Sealed this
Thirtieth Day of January, 1990**

Attest:

JEFFREY M. SAMUELS

Attesting Officer

Acting Commissioner of Patents and Trademarks